

Two-Lens and Lens-Fed Reflector Antenna Systems for MM-Wave Wireless Communications

Xidong Wu and George V. Eleftheriades

Department of Electrical and Computer Engineering
University of Toronto, 10 King's College Road, Ontario M5S 3G4, CANADA

1. Introduction

Recently the substrate lens antenna has been proposed for mm-wave broadband wireless communications [1-3]. The main advantages of this approach include highly directive antenna patterns, compatibility with IC techniques, suppression of surface-wave losses and the capability of multiple beam formation in a simple way. In [2], a realistic circularly polarized (CP) substrate lens antenna system operating at 30 GHz was demonstrated for single beam applications. Based on the same single lens antenna, a multiple beam circularly polarized antenna suitable for mm-wave broadband wireless communications was presented in [3]. However, both the reported single and multiple beam antennas employ a single lens which becomes bulky for high directivity applications at the low end of the mm-wave band.

In this paper, the characterization and design of 30 GHz two-lens and lens-fed reflector antenna systems is presented. In this approach, a hyperhemispherical substrate lens feeds either a larger but thin hyperbolic lens or a parabolic reflector. The attractive feature of these two-antenna systems is that it leads to a 75% reduction of the lens material, while maintaining about the same length and on-axis characteristics as the single lens antenna. In these 2-antenna systems, limited scan capability is possible which can lead to relaxed alignment requirements between a receiver and a transmitter for line-of-sight broadband wireless links.

2. Two-Lens Antenna

Figure 1a shows a schematic diagram of the two-lens antenna system. In this structure, a printed CP aperture coupled patch antenna [2] is used to excite a hyperhemispherical substrate lens which feeds a larger but thin hyperbolic objective lens. The system is focused when the hyperhemispherical lens is placed so that its virtual focus F coincides with the focal point of the hyperbolic lens f , as shown in Figure 1a. The lens is made out of Rexolite ($\epsilon_r=2.54$) which is easy to machine and leads to a reduced cost. The radiation patterns through the lens are computed using a geometrical-optics/diffraction theory approach [4].

The characterization strategy employed here is to determine the optimum f/D number for different radii R of the hyperhemispherical lens in terms of system efficiency. The diameter of the hyperbolic lens is fixed to $D=12\text{cm}$ aiming at a system directivity $D_{\text{sys}}=30\text{dB}$. The system directivity is calculated from $D_{\text{sys}}=4\pi U_{\text{max}}/P_{\text{rad}}$, where U_{max} is the maximum radiation intensity and P_{rad} is the power radiated by the printed patch. The corresponding system efficiency is defined as $\eta_{\text{sys}} = D_{\text{sys}} / (\pi D/\lambda_0)^2$, where λ_0 is the free space wavelength. The system efficiency can be decomposed according to $\eta_{\text{sys}} = \eta_1\eta_2\eta_s\eta_t$, where η_1 , η_2 are the

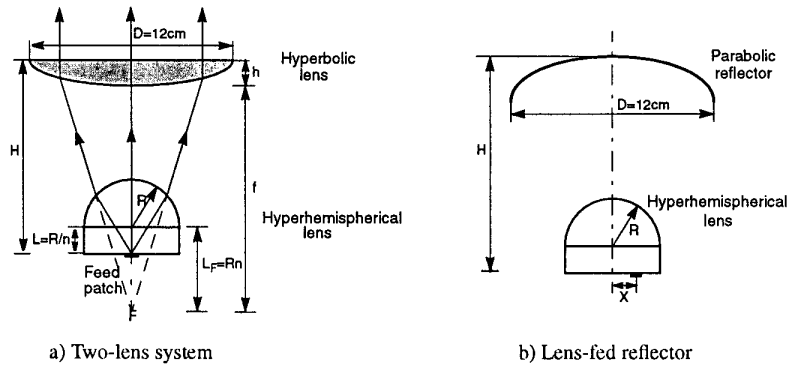


Figure 1. Schematic diagram of the two-antenna systems at 30 GHz

transmission efficiencies through the feed and objective lenses, η_s is the spillover efficiency and η_t is the taper efficiency on the objective lens. Figure 2 shows the simulated system efficiency and its breakdown as a function of the f/D number for three different feed lens radii. The transmission efficiency η_t is independent of the f/D number and is equal to 94.9% for various substrate lens radii R . As shown in Fig. 2, the optimum f/D numbers for $R=1.0\text{cm}$, 1.5cm and 2.0cm are about 0.90, 0.80 and 1.00, respectively. The corresponding system efficiencies are 43.2%, 43.8% and 43.1%. Among these three different radii, $R=1.5\text{cm}$ is the best choice since it results to a greater weight reduction than $R=2.0\text{cm}$, while maintaining almost the same system efficiency. Indeed, the optimized two-lens system can achieve 75% lens material reduction compared to a single lens solution of the same directivity [2], [3].

The two-lens antenna can also be used to launch multiple beams by printing an array at the back of the hyperhemispherical substrate lens. The scan angle depends on the off-axis displacement X/R where X is the off-axis distance, and R is the radius of the hyperhemispherical lens. Figure 3 shows the scan characteristics of the optimum focused two-lens antenna system with $R=1.5\text{cm}$ and $f/D=0.80$. The overall height H for the focused system is equal to 10.54cm . As shown, only limited scan angle, about 6.0° at X/R of 0.3, can be achieved for the focused system. One way to increase the scan capability is to reduce the distance between the substrate and objective lenses, leading to a defocused system. Figure 3 also shows the scan characteristics for two defocused systems with an overall height H of 9.14cm and 7.73cm . The scan angle is increased up to 9.6° at X/R of 0.3 for $H=7.73\text{cm}$. However, this is accompanied by a penalty of 2.0 dB reduction of the on-axis directivity. Similarly to the single lens antenna [3], there is no appreciable CP depolarization introduced in the two-lens system. In order to get a 3dB beam overlapping, the off-axis distance X for $H=7.73\text{cm}$ is found to be 0.36cm ($X/R = 0.24$). This implies that a patch array design with 3dB overlapping beams is possible since the maximum dimension of the used printed patches

at 30 GHz is 0.25cm [3]. With the requirement of a system directivity degradation smaller than 1.5dB from the peak value, which corresponds to an allowed largest off-axis displacement $X/R = 0.30$ (See Fig. 3, $H=7.73\text{cm}$), the substrate lens can house 7 patches in a hexagonal arrangement. The resulting 3dB scan coverage is about 16.8° , which can provide relaxed alignment requirements between a receiver and a transmitter for line-of-sight broadband wireless links.

3. Lens-Fed Reflector

The diagram of the lens-fed reflector is shown in Figure 1b, in which a reflector is used instead of an objective lens. This structure potentially provides improved system efficiency compared with the two-lens system, since the reflector introduces no reflection losses. In order to compare with the two-lens system, the diameter of the reflector D is also fixed at 12cm. Figure 4 shows the simulated system efficiency and its breakdown as a function of the f/D number for different substrate lens radii. As shown, the optimum f/D numbers for $R=1.0\text{cm}$, 1.5cm and 2.0cm are about 0.75, 0.83 and 0.92, respectively. The corresponding system efficiencies are 57.2%, 61.3% and 57.2%. Similarly to the two-lens system, the optimum radius of the hyperhemispherical lens occurs at $R=1.5\text{cm}$. From the efficiency analysis, both the two-lens system and the lens-fed reflector exhibit a similar taper efficiency. However, the two-lens system suffers from an additional 11% transmission loss ($\eta_2=89\%$), and demonstrates lower spillover efficiency ($\eta_s=68\%$) than that of the lens-fed reflector ($\eta_s=80\%$). Figure 5 shows the scan characteristics of the lens-fed reflector with $R=1.5\text{cm}$ and $f/D=0.83$ for $H=8.55\text{cm}$ (focused position), $H=7.36\text{cm}$ and $H=6.16\text{cm}$. The scan angle is about 11.0° at X/R of 0.3 for $H=6.16\text{cm}$.

4. Conclusions

Compared with a single substrate lens antenna, the described 2-antenna systems can achieve up to 75% reduction of the lens material while maintaining about the same length and on-axis characteristics. The lens-fed reflector provides higher system efficiency than the two-lens system. This makes the lens-fed reflector appealing for single beam applications. In these 2-antenna systems, limited scan capability with multiple beams cross-coupled at a 3dB level is possible, which can lead to relaxed alignment requirements between a receiver and a transmitter for line-of-sight broadband wireless links.

References

- [1] G.V. Eleftheriades, Y. Brand, J. Zürcher and J.R. Mosig, "ALPSS: A millimeter-wave aperture-coupled patch antenna on a substrate lens," *IEE Electronics Lett.*, vol. 33, No.3, pp. 169-170, Jan. 1997.
- [2] X. Wu, G.V. Eleftheriades, E. van Deventer, "A mm-wave circularly polarized substrate lens antenna for wireless communications," *Proceedings of the 1998 Antenn Conference*, Ottawa, pp. 595-598, August 1998.
- [3] X. Wu, G.V. Eleftheriades, E. van Deventer, "Design and characterization of single and multiple beam mm-wave circularly polarized substrate lens antennas for wireless communications," *Proc. of the IEEE Intl. Symposium on Antennas and Propagation*, pp. 2408-2411, Orlando, FL, July 1999.
- [4] D.F. Filipovic, S.S. Gearhart and G.M. Rebeiz, "Double-slot antennas on extended hemispherical and elliptical silicon dielectric lenses," *IEEE Trans Microwave Theory Tech.*, MTT-41, pp. 1738-1749, Oct. 1993.

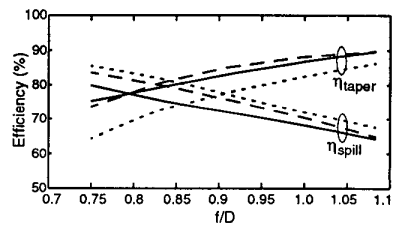
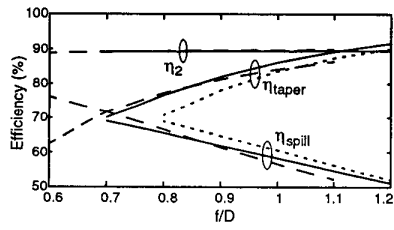
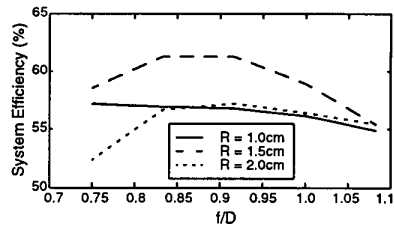
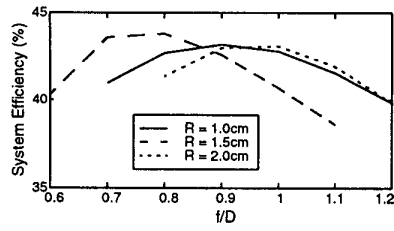


Figure 2. System efficiency of the focused two-lens system and its breakdown

Figure 4. System efficiency of the focused lens-fed reflector and its breakdown

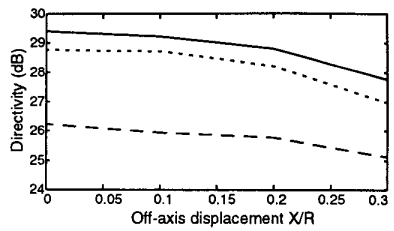
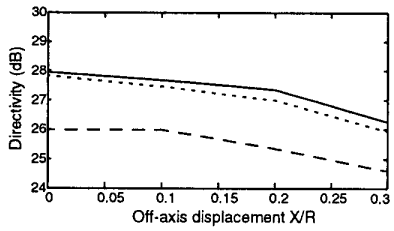
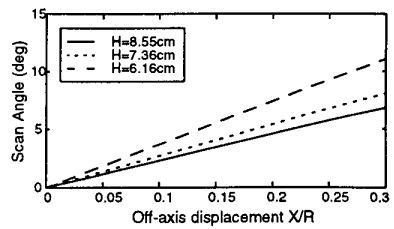
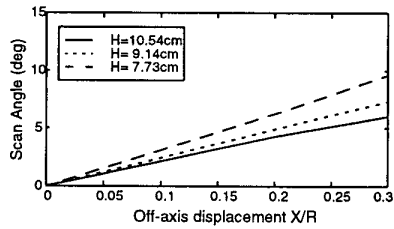


Figure 3. Scan characteristics of the two-lens antenna: $R=1.5\text{cm}$, $f/D=0.80$

Figure 5. Scan characteristics of the lens-fed antenna: $R=1.5\text{cm}$, $f/D=0.83$