

Assignment #1

ECE320F - Fields and Waves Fall 2003 / University of Toronto

1. Show that $e^{j\theta} = \cos \theta + j \sin \theta$.

$$\text{Recall that } e^{j\theta} = 1 + (j\theta) + \frac{(j\theta)^2}{2!} + \frac{(j\theta)^3}{3!} + \frac{(j\theta)^4}{4!} + \dots$$

$$\begin{aligned} \text{So } e^{j\theta} &= \left[1 + \frac{(j\theta)^2}{2!} + \frac{(j\theta)^4}{4!} + \dots \right] + \left[(j\theta) + \frac{(j\theta)^3}{3!} + \frac{(j\theta)^5}{5!} + \dots \right] \\ &= \left[1 - \frac{\theta^2}{2!} + \frac{\theta^4}{4!} + \dots \right] + j \left[\theta - \frac{\theta^3}{3!} + \frac{\theta^5}{5!} \right] \\ &= \cos \theta + j \sin \theta. \end{aligned}$$

2. The current through a circuit is found to be $\vec{I} = 6\angle -40^\circ$. What is the expression for the time dependent current with $\nu = 60$ Hz?

$$\begin{aligned} i(t) &= \operatorname{Re}\{\vec{I} e^{j\omega t}\} \\ &= \operatorname{Re}\{6\angle -40^\circ e^{j\omega t}\} \\ &= \operatorname{Re}\{6 e^{-j40^\circ} e^{j\omega t}\} \\ &= \operatorname{Re}\{6 e^{j(\omega t - 40^\circ)}\} \\ &= 6 \cos(\omega t - 40^\circ) \\ &= 6 \cos(2\pi \nu t - 40^\circ) \\ &= 6 \cos(2\pi(60 \text{ Hz})t - 40^\circ) \end{aligned}$$

$$i(t) = 6 \cos(120\pi t - 40^\circ)$$

Remember that the degree sign is important. If it is not present then units in radians are assumed. In this example an equivalent answer is

$$i(t) = 6 \cos\left(120\pi t - \frac{40}{180}\pi\right)$$

3. Prove that $\operatorname{Re}\{A\} = \frac{A + A^*}{2}$, where $A \in \mathbb{C}^2$.

$$\text{Let } A = a_1 + ja_2.$$

$$\text{Then } \operatorname{Re}\{A\} = a_1.$$

$$\text{RHS} = \frac{A + A^*}{2} = \frac{a_1 + \cancel{ja_2} + a_1 - \cancel{ja_2}}{2} = a_1$$

$$\text{LHS} = \text{RHS}.$$

4. Prove that for a sinusoidal function $a(z, t)$ where z is the space coordinate and t is the time, there exists a phasor $\bar{A}(z)$ such that

a. $\frac{\partial}{\partial t} a(z, t) = \frac{\partial}{\partial t} \operatorname{Re} \{ \bar{A}(z) e^{j\omega t} \}$ (1)

b. $\frac{\partial}{\partial t} a(z, t) = \operatorname{Re} \{ j\omega \bar{A}(z) e^{j\omega t} \}$. (2)

a). Let $a(z, t) = A \cos(\omega t + \phi(z))$.

Then
$$\begin{aligned} a(z, t) &= \operatorname{Re} \{ A \cos(\omega t + \phi(z)) + j A \sin(\omega t + \phi(z)) \} \\ &= \operatorname{Re} \{ A e^{j(\omega t + \phi(z))} \} \\ &= \operatorname{Re} \{ A e^{j\phi(z)} e^{j\omega t} \} \\ a(z, t) &= \operatorname{Re} \{ \bar{A}(z) e^{j\omega t} \}. \end{aligned}$$

Take derivative with respect to time of both sides

$$\boxed{\frac{\partial}{\partial t} a(z, t) = \frac{\partial}{\partial t} \operatorname{Re} \{ \bar{A}(z) e^{j\omega t} \}}$$

b). From LHS of (2)

$$\begin{aligned} \frac{\partial}{\partial t} a(z, t) &= \frac{\partial}{\partial t} [A \cos(\omega t + \phi(z))] \\ &= -A\omega \sin(\omega t + \phi(z)). \end{aligned}$$

From the RHS of (2)

$$\begin{aligned} \operatorname{Re} \{ j\omega \bar{A}(z) e^{j\omega t} \} &= \operatorname{Re} \{ j\omega A e^{j\phi(z)} e^{j\omega t} \} \\ &= \operatorname{Re} \{ j\omega A e^{j(\omega t + \phi(z))} \} \end{aligned}$$

$$= \operatorname{Re} \left\{ j\omega A [\cos(\omega t + \varphi(z)) + j \sin(\omega t + \varphi(z))] \right\}$$

$$= -\omega A \sin(\omega t + \varphi(z)).$$

Thus LHS = RHS and part (b) is proven.

This shows that the $\frac{\partial}{\partial t}$ operator can be moved inside the $\operatorname{Re}\{\}$, and that $\frac{\partial}{\partial t}$ can be substituted with $j\omega$ when only a single frequency ω is considered. ($\frac{\partial}{\partial t} \rightarrow j\omega$)

5. Show that $\frac{\partial^2}{\partial t^2} a(z, t) = \text{Re} \{ -\omega^2 \bar{A}(z) e^{j\omega t} \}$.

$$\begin{aligned} \text{Let } a(z, t) &= A \cos(\omega t + \varphi(z)) \\ &= \text{Re} \{ A e^{j\varphi(z)} e^{j\omega t} \}. \end{aligned}$$

$$\begin{aligned} \text{Then } \frac{\partial^2}{\partial t^2} a(z, t) &= \frac{\partial^2}{\partial t^2} \left[\text{Re} \{ A e^{j\varphi(z)} e^{j\omega t} \} \right] \\ &= \text{Re} \left\{ A e^{j\varphi(z)} \frac{\partial^2}{\partial t^2} e^{j\omega t} \right\} \\ &= \text{Re} \{ -\omega^2 A e^{j\varphi(z)} e^{j\omega t} \} \\ &= \text{Re} \{ -\omega^2 \bar{A}(z) e^{j\omega t} \}. \end{aligned}$$

6. Prove that if $\text{Re}\{\bar{A}(z)e^{j\omega t}\} = \text{Re}\{\bar{B}(z)e^{j\omega t}\}$, then $\bar{A}(z) = \bar{B}(z)$. This means that the $\text{Re}\{\}$ operator can be removed on phasors of the same frequency.

Since $\text{Re}\{\bar{A}(z)e^{j\omega t}\} = \text{Re}\{\bar{B}(z)e^{j\omega t}\}$ is true for all t , it is certainly true for $t=0$.

$\Rightarrow$ Set $t=0$.

$$\text{Re}\{\bar{A}(z)\} = \text{Re}\{\bar{B}(z)\}.$$

$$\text{Let } \bar{A}(z) = A e^{j\phi_A(z)} = A_r(z) + j A_i(z)$$

$$\bar{B}(z) = B e^{j\phi_B(z)} = B_r(z) + j B_i(z)$$

$$\text{Thus } \boxed{A_r(z) = B_r(z)}.$$

Now consider another specific time, $\omega t = \frac{\pi}{2}$. ($t = \frac{\pi}{2\omega}$)

$$\text{Re}\{\bar{A}(z)e^{j\frac{\pi}{2}}\} = \text{Re}\{\bar{B}(z)e^{j\frac{\pi}{2}}\}$$

$$\left(e^{j\frac{\pi}{2}} = j\right)$$

$$\text{Re}\{(A_r(z) + j A_i(z))j\} = \text{Re}\{(B_r(z) + j B_i(z))j\}$$

$$\text{Re}\{-A_i(z) + j A_r(z)\} = \text{Re}\{-B_i(z) + j B_r(z)\}$$

$$-A_i(z) = -B_i(z)$$

$$\boxed{A_i(z) = B_i(z)}$$

$$\text{Thus } A_r(z) + j A_i(z) = B_r(z) + j B_i(z)$$

$$\boxed{\bar{A}(z) = \bar{B}(z)}.$$

7. Show that $\operatorname{Re}\left\{\sum_{n=1}^M \overline{A_n}(z)\right\} = \sum_{n=1}^M \operatorname{Re}\{A_n(z)\}$, i.e. that the real part of the sum of phasors is the sum of the real parts.

$$\text{Let } \overline{A_n}(z) = A_{n,r}(z) + j A_{n,i}(z)$$

$$A_{n,r}(z) \in \mathbb{R}$$

$$A_{n,i}(z) \in \mathbb{R}$$

$$\begin{aligned} \text{Then } \operatorname{Re}\left\{\sum_{n=1}^M \overline{A_n}(z)\right\} &= \operatorname{Re}\left\{\sum_{n=1}^M [A_{n,r}(z) + j A_{n,i}(z)]\right\} \\ &= \operatorname{Re}\left\{\sum_{n=1}^M A_{n,r}(z) + j \sum_{n=1}^M A_{n,i}(z)\right\} \\ &= \operatorname{Re}\left\{\sum_{n=1}^M A_{n,r}(z)\right\} \\ &= \sum_{n=1}^M \operatorname{Re}\{A_{n,r}(z)\} \\ &= \sum_{n=1}^M \operatorname{Re}\{A_{n,r}(z) + j A_{n,i}(z)\} \\ &= \sum_{n=1}^M \operatorname{Re}\{\overline{A_n}(z)\} . \end{aligned}$$

8. Show that $\operatorname{Re}\{\bar{A}(z)\}\operatorname{Re}\{\bar{B}(z)\} = \frac{\operatorname{Re}\{\bar{A}(z)\bar{B}(z)\}}{2} + \frac{\operatorname{Re}\{\bar{A}(z)\bar{B}^*(z)\}}{2}$.

Start with RHS.

$$\begin{aligned}
 & \operatorname{Re}\left\{\frac{\bar{A}(z)\bar{B}(z)}{2}\right\} + \operatorname{Re}\left\{\frac{\bar{A}(z)\bar{B}^*(z)}{2}\right\} \\
 &= \frac{\operatorname{Re}\{[A_r(z) + jA_i(z)][B_r(z) + jB_i(z)]\}}{2} + \frac{\operatorname{Re}\{[A_r(z) + jA_i(z)][B_r(z) - jB_i(z)]\}}{2} \\
 &= \frac{\operatorname{Re}\{[A_r(z)B_r(z) - A_i(z)B_i(z)] + j[A_r(z)B_i(z) + A_i(z)B_r(z)]\}}{2} \\
 &\quad + \frac{\operatorname{Re}\{[A_r(z)B_r(z) + A_i(z)B_i(z)] + j[-A_r(z)B_i(z) + A_i(z)B_r(z)]\}}{2} \\
 &= \frac{A_r(z)B_r(z) - \cancel{A_i(z)B_i(z)}}{2} + \frac{A_r(z)B_r(z) + \cancel{A_i(z)B_i(z)}}{2} \\
 &= A_r(z)B_r(z) \\
 &= \operatorname{Re}\{\bar{A}(z)\} \cdot \operatorname{Re}\{\bar{B}(z)\}.
 \end{aligned}$$

Note that this proves that $\operatorname{Re}\{\bar{A}(z)\bar{B}(z)\} \neq \operatorname{Re}\{\bar{A}(z)\}\operatorname{Re}\{\bar{B}(z)\}$.

9. Show that $\langle \text{Re}\{\vec{A}(z)e^{j\omega t}\} \text{Re}\{\vec{B}(z)e^{j\omega t}\} \rangle_{\text{time avg}} = \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}^*(z)\}$.

Note that $\langle f(t) \rangle = \frac{1}{T_2 - T_1} \int_{T_1}^{T_2} f(t) dt$

where $[T_1, T_2]$ is the interval over which the time average is taken.

If $f(t)$ is periodic with period $T = T_2 - T_1$, we can take any interval, one of which is

$$\langle f(t) \rangle = \frac{1}{T} \int_0^T f(t) dt.$$

$$\langle \text{Re}\{\vec{A}(z)e^{j\omega t}\} \text{Re}\{\vec{B}(z)e^{j\omega t}\} \rangle = \frac{1}{T} \int_0^T \text{Re}\{\vec{A}(z)e^{j\omega t}\} \text{Re}\{\vec{B}(z)e^{j\omega t}\} dt$$

From the results of Problem 8,

$$= \frac{1}{T} \int_0^T \left[\frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}(z)e^{j2\omega t}\} + \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}(z)e^{j\omega t}e^{-j\omega t}\} \right] dt$$

$$= \frac{1}{T} \int_0^T \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}(z)e^{j2\omega t}\} dt + \frac{1}{T} \int_0^T \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}^*(z)\} dt$$

$$= \frac{1}{T} \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}(z) \int_0^T e^{j2\omega t} dt\} + \frac{1}{T} \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}^*(z) \int_0^T dt\}$$

$$= \frac{1}{T} \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}^*(z)\} \cdot T$$

$$= \frac{1}{2} \text{Re}\{\vec{A}(z)\vec{B}^*(z)\}.$$

Note that this gives

$$\langle \text{Power} \rangle = \langle v(t)i(t) \rangle = \frac{1}{2} \text{Re}\{\vec{V} \cdot \vec{I}^*\}.$$

10. The voltage across and current through a device are given by $v(t) = V_m \cos(\omega t + \theta_v)$ and $i(t) = I_m \cos(\omega t + \theta_i)$. What is the average power dissipated in the device?

$$v(t) = V_m \cos(\omega t + \theta_v) = \operatorname{Re}\{V_m e^{j\theta_v} e^{j\omega t}\} = \operatorname{Re}\{\bar{V} e^{j\omega t}\}$$

$$i(t) = I_m \cos(\omega t + \theta_i) = \operatorname{Re}\{I_m e^{j\theta_i} e^{j\omega t}\} = \operatorname{Re}\{\bar{I} e^{j\omega t}\}$$

Then the average power dissipated is

$$\langle P(t) \rangle = \langle v(t) i(t) \rangle = \langle \operatorname{Re}\{\bar{V} e^{j\omega t}\} \operatorname{Re}\{\bar{I} e^{j\omega t}\} \rangle$$

$$= \frac{1}{2} \operatorname{Re}\{\bar{V} \cdot \bar{I}^*\} \quad (\text{from Problem \#9})$$

$$= \frac{1}{2} \operatorname{Re}\{V_m e^{j\theta_v} I_m e^{-j\theta_i}\}$$

$$= \frac{1}{2} \operatorname{Re}\{V_m I_m e^{j(\theta_v - \theta_i)}\}$$

$$\boxed{\langle P(t) \rangle = \frac{1}{2} V_m I_m \cos(\theta_v - \theta_i)}$$

Resistors

$$v(t) = R i(t)$$

$$\bar{V} = R \bar{I} \Rightarrow \theta_v - \theta_i = 0$$

$$\langle P(t) \rangle = \frac{1}{2} V_m I_m = \frac{1}{2} \frac{V_m^2}{R} = \frac{1}{2} I_m^2 R$$

Note that this is different than $P_R = V \cdot I = \frac{V^2}{R} = I^2 R$.

To make a correspondence Let $V_{\text{eff}} = \frac{V_m}{\sqrt{2}}$, $I_{\text{eff}} = \frac{I_m}{\sqrt{2}}$

$$\text{Then } \langle P(t) \rangle = P_{\text{ave}} = \frac{V_m}{\sqrt{2}} \frac{I_m}{\sqrt{2}} = V_{\text{eff}} I_{\text{eff}}$$

V_{eff} also known as V_{RMS} (Root-Mean-Square).

Inductor

$$v(t) = L \frac{di(t)}{dt}$$

$$\bar{V} = L j\omega \bar{I}$$

$$V_m e^{j\theta_v} = L e^{j\frac{\pi}{2}} \omega I_m e^{j\theta_i}$$

$$\theta_v = \frac{\pi}{2} + \theta_i \Rightarrow \theta_v - \theta_i = \frac{\pi}{2}$$

$$\langle P(t) \rangle = \langle v(t) i(t) \rangle = \frac{1}{2} V_m I_m \cos\left(\frac{\pi}{2}\right) = 0.$$

Capacitor

$$i(t) = C \frac{dv(t)}{dt}$$

$$\bar{I} = C j\omega \bar{V}$$

$$I_m e^{j\theta_i} = C e^{j\frac{\pi}{2}} \omega V_m e^{j\theta_v}$$

$$\theta_i = \frac{\pi}{2} + \theta_v \Rightarrow \theta_v - \theta_i = -\frac{\pi}{2}$$

$$\langle P(t) \rangle = \langle v(t) i(t) \rangle = \frac{1}{2} V_m I_m \cos\left(-\frac{\pi}{2}\right) = 0.$$

Therefore the average power dissipated in an inductor or capacitor is zero. Note that the instantaneous power $p(t)$ is not in general zero, however.

11. A voltage source is connected to resistor R as shown in Fig. 1. The source produces the voltage waveform shown in Fig. 2.

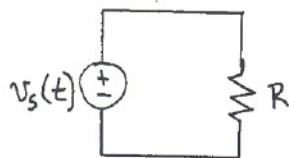


Figure 1.

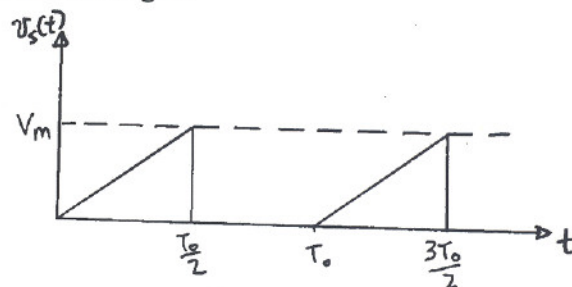


Figure 2.

- Find the instantaneous power delivered to R.
- What is the average power delivered to R?

$$v_s(t) = \begin{cases} \frac{2V_m}{T_0} t, & 0 \leq t \leq \frac{T_0}{2} \\ 0, & \frac{T_0}{2} < t < T_0 \end{cases}$$

$$v_s(t + T_0) = v_s(t).$$

$$a) \quad p(t) = \frac{v^2(t)}{R} = \begin{cases} \frac{1}{R} \left(\frac{2V_m}{T_0} t \right)^2, & 0 \leq t \leq \frac{T_0}{2} \\ 0, & \frac{T_0}{2} < t < T_0 \end{cases}$$

$$p(t + T_0) = p(t)$$

$$b) \quad P_{ave} = \langle p(t) \rangle = \frac{1}{T_0} \int_0^{T_0} p(t) dt$$

$$= \frac{1}{T_0} \int_0^{\frac{T_0}{2}} \frac{1}{R} \left(\frac{2V_m}{T_0} t \right)^2 dt$$

$$= \frac{1}{R} \frac{1}{T_0^3} 4V_m^2 \frac{1}{3} t^3 \Big|_0^{\frac{T_0}{2}}$$

$$= \frac{4V_m^2}{3RT_0} \cdot \frac{T_0^3}{8} = \boxed{\frac{V_m^2}{6R}}.$$

12. Show that the RMS (or effective value) for a sinusoidal voltage

$$v(t) = V_m \cos(\omega t + \theta) \text{ is } \frac{V_m}{\sqrt{2}}.$$

The RMS value is defined by

$$V_{RMS} = \sqrt{\frac{1}{T_0} \int_0^{T_0} v^2(t) dt}$$

$$= \sqrt{\frac{1}{T_0} \int_0^{T_0} V_m^2 \cos^2(\omega t + \theta) dt}$$

$$= \sqrt{\frac{1}{T_0} \int_0^{T_0} \left\{ \frac{V_m^2}{2} + \frac{V_m^2}{2} \cos[2(\omega t + \theta)] \right\} dt}$$

$$\left(\begin{array}{l} \text{Since} \\ \cos^2 x = \frac{1}{2} (1 + \cos 2x) \end{array} \right)$$

$$= \sqrt{\frac{1}{T_0} \left[\int_0^{T_0} \frac{V_m^2}{2} dt + \int_0^{T_0} \frac{V_m^2}{2} \cos[2(\omega t + \theta)] dt \right]}$$

$$= \sqrt{\frac{1}{T_0} \frac{V_m^2}{2} \int_0^{T_0} dt}$$

$$= \sqrt{\frac{1}{T_0} \frac{V_m^2}{2} T_0}$$

$$\boxed{V_{RMS} = \frac{V_m}{\sqrt{2}}}$$